Multi-level Thrust Ripples Minimization of Linear Flux Switching Motors With Segmented Secondary by Combined Genetic Algorithm and Response Surface Methodology

M. S. Hosseini*, H. Javadi* (C.A.) and S. Vaez-Zadeh**

Abstract: Linear flux switching motors with simple passive segmented secondary, referred as Segmented Secondary Linear Flux Switching Motors (SSLFSMs), have low cost secondary and therefore are applicable to transportation systems like Maglev. However, it is shown that the SSLFSMs suffer from high thrust ripples. In this paper, minimizing SSLFSM thrust ripples besides maximizing its developed thrust are performed by considering the motor dimensions as design variables. Since the optimization of the motor is a high dimensional problem, a multi-level optimization method is employed to improve the machine performances and efficiency. According to the effects of the design variables on the optimization objectives, a sensitivity analysis is carried out to divide the design variables into two levels: mild-sensitive level and strong-sensitive level. Then, the two levels of design variables are optimized based on a mathematical model. Two different optimization methods as the Design of Experiment (DOE) and the Response Surface Method (RSM) are used in mild-sensitive level and the Genetic Algorithm (GA) is also used in strong-sensitive level. Based on FEM analysis, electromagnetic performance of the original motor and the optimal one are compared and the validity of the proposed optimization method is verified. Also, the effectiveness of the mathematical model used in thrust and thrust ripples calculations is evaluated and verified.

Keywords: Linear flux Switching Motors, Multi-Level Optimization, Response Surface Methodology, Sensitivity Analysis Method, Thrust Ripples Minimization.

1 Introduction

LINEAR Flux Switching Motors (LFSMs) inherit most merits of rotary Flux Switching Motors (FSMs) that leads to simple and robust structure of secondary and high force density [1-2]. When the stator is very long, e.g. in the case of Maglev or underground train drives, such a simple and low cost secondary is attractive. Recently, a LFSM with a novel structure, referred as Segmented Secondary Linear Flux Switching Motor (SSLFSM), is introduced for rail transportation systems [3]. In such a structure, the armature and field windings are wounded in the primary slots and the secondary is composed of simple laminated segments. The power supply for DC excitation can be placed on the mover and the AC currents of armature windings can be drawn from transmission lines above the train and along the track. Bi-polar flux linkage [4] leading to high force density and close to sinusoidal back-emf facilitating less complicated control strategy [5-6] are other advantages of FSMs that are inherited to SSLFSM. Improved sinusoidal back-emf of SSLFSMs is achieved by some modifications into their structure [7]. Nevertheless, it has higher thrust ripples caused by the segmented structure of secondary that should be minimized. A d-q model representation of SSLFSMs is achieved based on Finite Element Method (FEM) [8].
but thrust ripples calculation is not presented.

Optimizing average thrust of different types of linear flux switching motors are studied based on FEM [9-10]. Cogging torque reduction of rotary flux switching motors [11-13] beside its average torque improvement [12-13] are studied by FEM analysis too. However, the FEM-based method is time consuming and needs intensive computations. Thrust [14] and thrust ripple minimization [15] of LSMs based on an analytical layer modeling are studied in the literature. Layer modeling is fast against FEM-based modeling, but it is less accurate especially in the case of motors with high saliency in primary and secondary as SSLFSMs.

Here a mathematical model to calculate the motor thrust and thrust ripples is recalled [16]. This mathematical model is necessary for design optimization of the newly developed SSLFSMs and is fast and accurate enough in comparison with FEM-based and layer modeling methods. Besides, minimizing the thrust ripples (which is the main aim of this paper) and maximizing average thrust, which is very important in propulsion systems are other objectives of the optimization problem. As a result, a dual-objective optimization is carried out. The motor dimensions are considered as design variables. According to the effect of the design variables on the optimization objectives, sensitivity analysis method is used to divide the design variables into two levels: mild-sensitive level and strong-sensitive level. Multi-level optimization procedure is effective in the case of high dimensional optimization problems [12-13], [17]. By appropriate braking of the problem into different levels, the optimization time and accuracy will be improved. Sensitivity analysis is a common method in dividing the optimization levels [12-13, 17].

Since SSLFSM is a new motor, its optimization is absent in the literature. In this paper, it is tried to improve machine characteristics. Also, multi-level optimization for this motor is done for the first time. In addition, the analysis method beside the multilevel optimization of the SSLFSM is novel and can be applied to other motors. Also, the newly developed machine model [16] is adapted to the multi-level sensitive analysis based optimization procedure in this paper.

Different sections of the paper are as following. Motor topology and its 2D analytical model are presented in Section 2. The machine are divided into different regions and an exact solution is obtained for each region by solving Maxwell and Poisson equations. Optimization problem considering the developed thrust and thrust ripples as optimization objectives and different dimensions of the motor as design variables, are defined in Section 3. Then, the design optimization of the motor based on the analytical model is carried out in Section 4. Sensitivity analysis is performed to divide the design variables into two levels: mild-sensitive level and strong-sensitive level. The Design of Experiment (DOE) and the Response Surface Method (RSM) is used as optimization method in mild-sensitive level and Genetic Algorithm (GA) in strong-sensitive level. Design evaluation based on a finite element analysis is carried out in Section 5. The effectiveness of the mathematical model used in thrust and thrust ripples calculations and the validity of the proposed optimization method are studied. Finally, some conclusions are presented in Section 6.

2 Machine Model

2.1 Motor Topology and Operation Principles

Fig. 1 shows a schematic view of a single-sided SSLFSM with a long secondary [18]. The primary, which is the mover of the motor, includes both three-phase winding coils and DC excitation field coils alternatively. The polarization of each excitation winding is opposite to the next one. The secondary of SSLFSM is composed of simple segments which makes this topology applicable to rail transportation systems.

DC windings and secondary segments constitute electromagnetic poles. When the mover moves along the secondary and the field coils are excited, the polarity of the armature winding flux linkage changes as well as its value. This bipolar flux linkage induces a sinusoidal back-emf in the armature coils. Nevertheless, salient structure of the motor primary and secondary causes the

Fig. 1 Parameters and dimensions of the SSLFSM. 
(a) 3-dimensional view of the motor and 
(b) 2-dimensional view of the motor illustrating detailed structure of segments and primary teeth.
generation of undesirable thrust ripples.

The parameters and dimensions of the motor are as follow: \( L \) is the motor width along the \( z \) direction, \( g \) is the airgap length, \( r_\tau \) is twice the motor pole pitch, which is the distance between one segment to the next one, \( w_{seg} \) is the segment width, \( w_{frnt} \) is the segment front part width, which is in front of air gap, \( w_{back} \) is the segment back part width, which lies on the track, \( h_{seg} \) is the segment height, \( w_s \) is the slot width, \( h_s \) is the slot height, \( w_a \) is the slot opening height, \( w_f \) is the field tooth width, \( w_{arm} \) is the armature tooth width, \( \gamma \) is the tooth tip angle.

2.2 Analytical Model [16]

In this subsection a method of developing an analytical model for the SSLFSM is briefly explained [19]. The method is based on solving Maxwell and Poison equations in certain regions of the motor shown in Fig. 2. It is known that the magnetic flux density (\( B \)) in each region can be written as:

\[
\vec{B} = \nabla \times \vec{A}
\]  
(1)

where \( \vec{A} \) is the magnetic vector potential. The magnetostatic Maxwell equations lead to a Poisson equation, given by:

\[
\nabla^2 \vec{A} = -\mu_0 \mu \vec{J}
\]  
(2)

where \( \mu_0 \) is the relative permeability of each region. Since the current density vector \( \vec{J} \) has only a component in the \( z \)-direction, \( \vec{A} \) also has a component in \( z \)-direction, therefore, Eq. (2) reduces to:

\[
\frac{\partial^2 A_z(x, y)}{\partial x^2} + \frac{\partial^2 A_z(x, y)}{\partial y^2} = -\mu_0 \mu J_z
\]  
(3)

In order to solve (3) in the 12 slot regions, it is necessary to describe current sources in term of Fourier series. Current source description is obtained by applying the imaging method [19] as:

\[
J_i = J_{i0} + \sum_{s=1}^{2} \left( J_{i1} \sin(\omega_s x_i) + J_{i2} \cos(\omega_s x_i) \right)
\]  
(4)

\[
\begin{align*}
J_{i0} &= \frac{1}{2r_s} \int_{-r_s}^{r_s} J_0 \, dx_n = \frac{J_{i0} + J_{i2}}{2} \\
J_{i1}(n) &= \frac{1}{r_s} \int_{-r_s}^{r_s} J_1 \sin(\omega_s x_i) \, dx_n = 0 \\
J_{i2}(n) &= \frac{1}{r_s} \int_{-r_s}^{r_s} J_2 \cos(\omega_s x_i) \, dx_n \\
&= \frac{2}{n\pi} \sin \frac{n\pi}{2} \left( J_{i1} + (-1)^n J_{i2} \right)
\end{align*}
\]  
(5)

where \( J_{i0} \) and \( J_{i2} \) are current density of half area in \( i \)-th slot term in slot of \( A/n^2 \), \( r_s \) is the width of the slot regions, S1-S12, and \( \omega_s \) is the spatial frequency of the region given by:

\[
\omega_s(n) = \frac{n\pi}{r_s}
\]  
(6)

The solution for the magnetic vector potential is written in terms of Fourier components as the following:

\[
\vec{A} = A_i(x_i, y) \hat{z}
\]

\[
= \left[ A_{i0} + \sum_{s=1}^{2} \left( A_{i1}(y) \sin(\omega_s x_i) + A_{i2}(y) \cos(\omega_s x_i) \right) \right] \hat{z}
\]  
(7)

where \( \omega_s \) is the spatial frequency of the region \( k \), defined as:

\[
\omega_s = \frac{n\pi}{r_s}
\]  
(8)

Hence, the expressions for the magnetic flux density distribution in different regions are given as:

\[
\vec{B} = \left( \sum_{s=1}^{2} \left( B_{i0} \sin(\omega_s x_i) + B_{i2} \cos(\omega_s x_i) \right) \right) \hat{x} \\
+ \left( \sum_{s=1}^{2} \left( B_{i1}(y) \sin(\omega_s x_i) + B_{i2}(y) \cos(\omega_s x_i) \right) \right) \hat{y}
\]  
(9)

where the functions \( B_{i0}, B_{i1}, B_{i2} \) and \( B_{i2} \) can be obtained by considering the transfer relations for every coordinate system [19] and are given by:

\[
\begin{align*}
B_{i0} &= a_e e^{\omega_m t} + b_e e^{-\omega_m t} + \frac{\mu J_s}{\omega_s} \\
B_{i1} &= c_e e^{\omega_m t} - d_e e^{-\omega_m t} - \frac{\mu J_s}{\omega_s} \\
B_{i2} &= c_e e^{\omega_m t} - d_e e^{-\omega_m t} \\
B_{i2} &= -\mu J_{i1} x + B_{i0}
\end{align*}
\]  
(10)

The set of unknowns \( a_e, b_e, c_e, d_e \) and \( B_0 \) for every region is obtained by solving (10), considering the Neumann and continuous boundary conditions in the \( y \)-direction. As a result, the normal component of flux density at \( y=y_0 \) in the airgap is given by (11):

\[
\begin{align*}
B_z(x, y) &= \sum_{s=1}^{2} \left[ \sin(\omega_s x) \left( a_{s0} e^{\omega_s t} + b_{s0} e^{-\omega_s t} \right) \\
&+ \cos(\omega_s x) \left( -c_{s0} e^{\omega_s t} - d_{s0} e^{-\omega_s t} \right) \right]
\end{align*}
\]  
(11)

where \( \omega_s \) is the spatial frequency of the region \( g \), \( n_e \) is the number of harmonics in region \( g \).

3.3 Thrust and Thrust Ripples Calculations

In order to calculate the motor thrust, virtual work method is employed, while for the sake of
simplification, the stored electromagnetic energy in the iron part of the motor is considered unchanged. Electromagnetic energy stored in the airgap \( W_{c(y)} \) is calculated for different positions of the motor as (12) [20]:

\[
W(x,y) = \int_{0}^{M} B_{c}^{2}(x',y) \frac{2}{\mu_{0}} \times L \times g \times dx' \quad (12)
\]

where \( M \) is the primary total length and \( B_{c}(x,y) \) is the airgap flux density produced by both field excitation and armature currents in different positions of \( x \) and \( y \). Differential variation of energy along the \( x \) direction equals to the detent force obtained as (13):

\[
F(x) = \frac{dW(x,y)}{dx} = \frac{d}{dx} \left( \int_{0}^{M} B_{c}^{2}(x',y) \frac{2}{\mu_{0}} \times L \times g \times dx' \right) \quad (13)
\]

It should be mentioned that in the case of small airgap, the flux density variation in the airgap along \( y \) axis is not considerable. Therefore, middle of the airgap \( y = g/2 \) is assumed as a reference point. Thrust ripple equation is as (14):

\[
\text{Thrust - ripple} = \frac{\text{max} - \text{min}}{2 \times \text{ave}} \times 100
\quad (14)
\]

In this analysis thrust ripples caused by salient structure of the motor is considered, whereas the end effect related force ripples are ignored. Has been shown that, the end effect of the motor is negligible and independent of its moving speed [8]. This is due to the fact that the airgap flux density in the case of short primary with end teeth or long primary are the same [8].

3 Optimization Problem

An optimization problem with \( p \) objectives, \( n \) variables, and \( m \) constraints is formulated as:

Maximize \( f_{1}(x), f_{2}(x), \ldots, f_{p}(x) \), \( x \in K \) \quad (15)

where \( x \in \mathbb{R}^{n} \), \( f: \mathbb{R}^{n} \rightarrow \mathbb{R} \). Also, \( K \) is a feasible set of solution (15) which is described by (16):

\[
K = \{ x \in \mathbb{R}^{n} : g_{i}(x) \leq 0, i = 1,2,\ldots,p \} \quad (16)
\]

where \( g_{i}(x) \) are optimization constraints which limit the design variables. The design variables are chosen as segment height, segment factor \( (s_{f}) \) defined as \( w_{seg}/f_{p} \), segment back factor \( (s_{bf}) \) defined as \( w_{bf}/w_{seg} \), Segment Chamfer Length Ratio (SCLR), defined as \( w_{bf}/w_{seg} \), field tooth width, armature tooth width, slot width, slot height, slot opening width, slot opening height, and tooth tip angle based on their important influence on the design optimization as considered later in this section. The design objectives in this paper are to maximize the motor developed thrust and minimize its thrust ripples. A 12/8 SSLFSM of the type depicted in Fig. 1 is selected as the basis for optimization. Figs. 3-5 show the variations of objective functions in terms of motor dimensions. Figs. 3(a) and (b) show variations of motor average thrust and thrust ripples in terms of armature and field teeth widths, respectively, when the other motor variables are constant. As it is seen, both objectives increase with an increase in the armature tooth width. Also, field tooth width has a strong effect on both the motor average thrust and thrust ripples. It can be seen that the objectives do not have simple common optimal point as far as teeth widths are concerned. Therefore, both design variables should be considered in the optimization procedure. Figs. 3(c) and (d) show variations of the motor average thrust and thrust ripples in terms of slot width and height. It is seen that the motor average thrust increases and its thrust ripples decrease with a decrease in the slot height. Therefore, to maximize the developed thrust of the motor and minimize its ripples, the slot height should be set to its minimum value. However, changes in slot width strongly affects both objectives, but it does not have specific rhythm. Therefore, the design variables of slot width should be considered in the optimization procedure to find the optimal point. Figs. 4(a) - (d) show variations of two objectives in terms of slot opening width and height, segment back factor and segment height, respectively. It is seen that the motor thrust ripples increase with an increase in slot opening height and its average thrust decreases with an increase in segment back factor. As it is seen in Figs. 4(a) - (d), the
The motor average thrust and thrust ripples variations in terms of SCLR, tooth tip angle and decrease in tooth tip angle and an increase in segment factor, leads to an increase in the motor segment factor are shown in Figs. 5(a)-(c), respectively. It is seen that an increase in SCLR, an average thrust. However, changes in these four variables of slot opening width and height, segment back factor and segment height strongly affect the two objectives but without any specific rule. As a result, to find the optimal point of the optimization problem, it is necessary to consider these design variables as optimization variables.
three design variables do not lead to a certain variation in the motor thrust ripples. Therefore, these design variables should also be taken into account in the optimization procedure. In order to prevent unrealistic optimization results, some constraints are applied to the design variables which are presented in Table 1. As it can be seen a minimum and a maximum value are assigned for each design variable.

### Design Optimization

Since the optimization problem is a high dimensional one, a two-level optimization method is efficient to improve the performances as well as the optimization efficiency. Two-level optimization method leads to more accuracy and less computational time. So, in this section, a sensitivity method is adopted and two level design variables are defined: mild-sensitive level and strong-sensitive level. Then, to improve the whole design efficiency, the DOE &RSM and GA are applied in the mild-sensitive level and strong-sensitive level, respectively. Also, other optimization method can be used but here DOE & RSM and GA methods are chosen [12]. Fig. 6 shows the flowchart of the total optimization process.

#### 4.1 Sensitivity Analysis Method

In order to break a large scale optimization problem into small parts, it is necessary to evaluate each design variable effect on the optimization objectives individually and independently. To achieve this, sensitive analysis method is useful and effective. In this

### Table 1 Design constraint.

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Unit</th>
<th>Symbol</th>
<th>Min</th>
<th>Max</th>
</tr>
</thead>
<tbody>
<tr>
<td>Field tooth width</td>
<td>mm</td>
<td>$w_{tf}$</td>
<td>14</td>
<td>50</td>
</tr>
<tr>
<td>Armature tooth width</td>
<td>mm</td>
<td>$w_{ta}$</td>
<td>13</td>
<td>20</td>
</tr>
<tr>
<td>Slot width</td>
<td>mm</td>
<td>$w_s$</td>
<td>14</td>
<td>23</td>
</tr>
<tr>
<td>Slot height</td>
<td>mm</td>
<td>$w_o$</td>
<td>40</td>
<td>75</td>
</tr>
<tr>
<td>Slot opening width</td>
<td>mm</td>
<td>$w_{o1}$</td>
<td>8</td>
<td>16</td>
</tr>
<tr>
<td>Slot opening height</td>
<td>mm</td>
<td>$w_{o2}$</td>
<td>1</td>
<td>9</td>
</tr>
<tr>
<td>Segment height</td>
<td>mm</td>
<td>$h_s$</td>
<td>12</td>
<td>20</td>
</tr>
<tr>
<td>Segment back-factor</td>
<td>-</td>
<td>$w_{se}$</td>
<td>0.4</td>
<td>1.2</td>
</tr>
<tr>
<td>Segment factor</td>
<td>-</td>
<td>$w_{se}$</td>
<td>0.5</td>
<td>0.9</td>
</tr>
<tr>
<td>Tooth tip angle</td>
<td>deg</td>
<td>$h_m$</td>
<td>5</td>
<td>40</td>
</tr>
<tr>
<td>SCLR</td>
<td></td>
<td>SCLR</td>
<td>0.1</td>
<td>0.9</td>
</tr>
</tbody>
</table>

Fig. 5 Thrust and thrust ripples variations in terms a) SCLR, b) tooth tip angle, and c) segment factor.

Fig. 6 The total flowchart of optimization procedure.
way a sensitivity index \( (H(x_i)) \) considering constraints of Table 1 is chosen for each design variable as following [13]:

\[
H(x_i) = \frac{V(E(y/x_i))}{V(y)}
\]  

(17)

where \( x_i \) is the design variable, \( y \) is optimization objectives, \( E(y/x_i) \) is the average value of \( y \) when \( x_i \) is constant at its initial value and other design variables changes from their minimum value to their maximum value, \( V(E(y,x_i)) \) is the variance of \( E(y,x_i) \), and \( V(y) \) is variance of \( y \). Since there are two optimization objectives, a comprehensive sensitivity function \( S(x_i) \) which combines the sensitivity of two objectives into a single one is introduced as following [12], [13]:

\[
S(x_i) = \lambda_1 H_{thrust}(x_i) + \lambda_2 H_{ripple}(x_i)
\]

(18)

where \( \lambda_1 \) and \( \lambda_2 \) are the weight coefficients of the developed thrust and thrust ripples, which satisfy the equation \( \lambda_1 + \lambda_2 = 1 \); \( H_{thrust}(x_i) \) and \( H_{ripple}(x_i) \) are the sensitivity indices of the two objectives. Since the importance of the two optimization objectives in this paper are the same, \( \lambda_1 \) and \( \lambda_2 \) are set to 0.5. Then the sensitivity function of each design variable is easily obtained from Eqs. (17) and (18). Table 2 illustrates the sensitivity indices and the comprehensive sensitivity function values for each design variable. In the case of selecting a separating condition at \( S(x_i) < 0.04 \), half of the design variables have comprehensive sensitivity function value more than this threshold and the other half have comprehensive sensitivity function value less than 0.04. Therefore, the separating condition is chosen considering comprehensive sensitivity function value for different design variables. In the case of \( S(x_i) > 0.04 \), the design variable of \( x_i \) is optimized in strong sensitive level (i.e. level 1) and otherwise is optimized in the mild-sensitive level (i.e. level 2). In this way the design variables are classified into two different groups and consequently the initial optimization problem is broken into two smaller problems.

4.2 GA in Strong-Sensitive Level Optimization

Considering the results of the sensitivity function presented in Table 2, \( S(x_i) \) is bigger than 0.04 for design variables of field tooth width, slot width, segment back factor, segment factor and SCLR. Therefore, these five variables are selected to be optimized in the strong-sensitive level. GA is chosen as optimization algorithm of this level. Since there are two optimization objectives, a weighted function with the same importance for each objective is employed as following [21]:

\[
f = 0.5 \times f_{thrust} + 0.5 \times f_{ripple}
\]

(19)

Therefore maximizing the function \( f \) is the single optimization objective of the GA. Since the weight of each objective (thrust and thrust ripple) are the same at 0.5, the algorithm tends to optimize them in the same order [21]. In order to achieve to the optimal point, an initial population with 25 population size is randomly produced. The best answer of this population is directly transferred to the next generation. Selection, crossover and mutation as main operators of GA are used to produce other answers of each generation.

In this paper, the selection operator employs the Tournament and the Roulette wheel methods with equal probabilities and the probability for the crossover and mutation operators equal 0.93 and 0.07, respectively [22]. It should be mentioned that 40 iterations are carried out to reach the optimal point step by step.

4.3 DOE and RSM in Mild-Sensitive Level Optimization

Considering the results of sensitivity analysis presented in Table 2, \( S(x_i) \) is less than 0.04 for design variables of armature tooth width, slot opening width, slot opening height, segment height and tooth tip angle which are respectively coded as \( X_1, X_2, X_3, X_4 \) and \( X_5 \). So, these variables have less effect on optimization objectives compared to other variables. As a result they are considered as a group to be optimized in mild-sensitive level. In this level DOE and RSM are chosen as optimization methodology. The value of each design variable is normalized in interval of [-1, 1]. The two objectives of maximizing the motor average thrust and minimizing its thrust ripples are coded as \( Y_1 \) and \( Y_2 \) respectively. It is tried to find the RSM common second order models for \( Y_1 \) and \( Y_2 \) in terms of \( X_1-X_5 \) as following [23]:

\[
Y_i = \alpha_i + \beta_i X_1 + \gamma_i X_2 + \delta_i X_3 + \epsilon_i X_4 + \zeta_i X_5 + \\
\eta_i X_1 X_1 + \upsilon_i X_2 X_2 + \xi_i X_3 X_3 + \rho_i X_4 X_4 + \varsigma_i X_5 X_5 + \\
\sigma_i X_1 X_2 + \chi_i X_1 X_3 + \upsilon_i X_2 X_1 + \rho_i X_3 X_3 + \delta_i X_5 X_5 + \xi_i X_2 X_3 \times X_4 \\
X_i + \phi_i X_5 X_5 \quad \text{for } i = 1, 2
\]

(20)

To achieve this, a DOE with a \( 2^{5-2} = 8 \) fractional factorial Central Composite Design (CCD) is

<table>
<thead>
<tr>
<th>Parameter</th>
<th>( H_{thrust} )</th>
<th>( H_{ripple} )</th>
<th>( S(x_i) )</th>
</tr>
</thead>
<tbody>
<tr>
<td>Field tooth width</td>
<td>0.683</td>
<td>0.925</td>
<td>0.804</td>
</tr>
<tr>
<td>Armature tooth width</td>
<td>0.015</td>
<td>0.058</td>
<td>0.037</td>
</tr>
<tr>
<td>Slot width</td>
<td>0.012</td>
<td>0.077</td>
<td>0.044</td>
</tr>
<tr>
<td>Slot opening width</td>
<td>0.011</td>
<td>0.006</td>
<td>0.009</td>
</tr>
<tr>
<td>Slot opening height</td>
<td>0.011</td>
<td>0.031</td>
<td>0.021</td>
</tr>
<tr>
<td>Segment height</td>
<td>0.010</td>
<td>0.0002</td>
<td>0.005</td>
</tr>
<tr>
<td>Segment back-factor</td>
<td>0.317</td>
<td>0.001</td>
<td>0.159</td>
</tr>
<tr>
<td>Segment factor</td>
<td>0.72</td>
<td>0.101</td>
<td>0.410</td>
</tr>
<tr>
<td>Tooth tip angle</td>
<td>0.0009</td>
<td>0.00005</td>
<td>0.0005</td>
</tr>
<tr>
<td>SCLR</td>
<td>0.218</td>
<td>0.132</td>
<td>0.175</td>
</tr>
</tbody>
</table>
employed [24]. Also, 4 center points of (0,0,0,0,0), and 10 axial points of (+1,0,0,0,0), (0,+1,0,0,0), ... and (0,0,0,0,+1) are selected. Totally, the results of 22 runs are presented in Table 3. Then, the coefficients of the second order models of (10) are obtained with Minitab Software. These results are presented in Table 4. In order to solve a two-objective (Y1 and Y2) problem, a weighted additive Goal Programming (GP) method is employed as (21) and (22):

\[ \text{max } Z = \sum_{j=1}^{2} w_j \alpha_j \quad (21) \]

\[ w_j = 0.5 \]

\[ \alpha_j \in [0,1] \quad \text{for } j = 1,2 \]

\[ \alpha_j \leq \mu_{\alpha_j} \quad (22) \]

where \( \alpha_j \) and \( w_j \) denote the achievement degree and the weight of \( j \)-th goal, respectively. The \( \mu_{\alpha_1} \) and \( \mu_{\alpha_2} \) are membership functions of the two objectives that are respectively given as (23) and (24). These functions are obtained based on the results of a pay-off table of a Positive Ideal Solution (PIS) as presented in Table 5 [10].

\[ \mu_{\alpha_i} = \begin{cases} 0 & Y_i \leq 0.54 \\ Y_i - 0.54 & 0.54 < Y_i < 1.02 \\ 1 & Y_i \geq 1.02 \end{cases} \quad (23) \]

\[ \mu_{\alpha_i} = \begin{cases} 0 & Y_i \leq 0.91 \\ 1.04 - 0.91 & 0.91 < Y_i < 1.04 \\ 1 & Y_i \geq 1.04 \end{cases} \quad (24) \]

The presented model is solved with Lingo software and the optimal point of the mild-sensitive level is obtained. The optimization results for the SSLFSM considering all design variables in strong-sensitive level or in mild-sensitive level, are presented in Table 6 and are compared with the initial quantity values of motor. As it is seen, the motor average thrust is increased considerably from 599N to 1064N and its thrust ripples is reduced from 78% to 96%.

5 Design Evaluation

The design optimizations in level 1 and 2 of this work were carried out based on the analytical machine model presented in section II. So it is necessary to evaluate the model accuracy. A 2D finite element analysis is employed to evaluate the presented model. Fig. 7(a) shows the FEM model for the initial SSLFSM. Fig. 7(b) illustrates and compares thrust and thrust ripples obtained from the analytical model and 2D FEM analysis for the initial motor. These results are numerically compared in Table 6 which can be seen, the initial motor average thrust obtained from 2D FEM and the analytical model is respectively 578.5N and 599N, i.e. with 3.54% error. Also, thrust ripples obtained from 2D FEM analysis and the analytical model are respectively 77.6% and 78% with 0.52% difference. Therefore, there is good agreement between the analytical model and FEM analysis results both in the values and the shapes development.

Fig. 8(a) shows the FEM model for the optimized motor. Compared to the initial motor, the segments and the teeth dimensions have been totally changed. This optimization leads to a very smoother and stronger motor thrust. Fig. 8(b) illustrates and compares thrust and thrust ripples of the optimized motor obtained from the analytical model and 2D FEM. The optimized motor

<table>
<thead>
<tr>
<th>Input Variables</th>
<th>Output Variables</th>
</tr>
</thead>
<tbody>
<tr>
<td>Run</td>
<td>X1</td>
</tr>
<tr>
<td>1</td>
<td>-1</td>
</tr>
<tr>
<td>2</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>-1</td>
</tr>
<tr>
<td>4</td>
<td>1</td>
</tr>
<tr>
<td>5</td>
<td>0</td>
</tr>
<tr>
<td>6</td>
<td>-1</td>
</tr>
<tr>
<td>7</td>
<td>0</td>
</tr>
<tr>
<td>8</td>
<td>0</td>
</tr>
<tr>
<td>9</td>
<td>0</td>
</tr>
<tr>
<td>10</td>
<td>1</td>
</tr>
<tr>
<td>11</td>
<td>1</td>
</tr>
<tr>
<td>12</td>
<td>1</td>
</tr>
<tr>
<td>13</td>
<td>0</td>
</tr>
<tr>
<td>14</td>
<td>1</td>
</tr>
<tr>
<td>15</td>
<td>0</td>
</tr>
<tr>
<td>16</td>
<td>-1</td>
</tr>
<tr>
<td>17</td>
<td>0</td>
</tr>
<tr>
<td>18</td>
<td>0</td>
</tr>
<tr>
<td>19</td>
<td>-1</td>
</tr>
<tr>
<td>20</td>
<td>0</td>
</tr>
<tr>
<td>21</td>
<td>0</td>
</tr>
<tr>
<td>22</td>
<td>0</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Coefficient</th>
<th>Coefficient</th>
<th>( r=1 )</th>
<th>( r=2 )</th>
<th>Coefficient</th>
<th>Coefficient</th>
<th>( r=1 )</th>
<th>( r=2 )</th>
</tr>
</thead>
<tbody>
<tr>
<td>( a_1 )</td>
<td>0.99525</td>
<td>0.7632</td>
<td>( \rho_1 )</td>
<td>-0.0254</td>
<td>-0.0809</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \beta_1 )</td>
<td>0.0831</td>
<td>0.0131</td>
<td>( \alpha_1 )</td>
<td>-0.0001</td>
<td>-0.0056</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \lambda_1 )</td>
<td>0.0495</td>
<td>0.0979</td>
<td>( \eta_1 )</td>
<td>0.0052</td>
<td>0.0327</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \delta_1 )</td>
<td>-0.0324</td>
<td>-0.1341</td>
<td>( \delta_1 )</td>
<td>0.0159</td>
<td>0.0069</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \omega_1 )</td>
<td>0.0219</td>
<td>0.0431</td>
<td>( \omega_1 )</td>
<td>0.0169</td>
<td>0.111</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \zeta_1 )</td>
<td>-0.0057</td>
<td>-0.0041</td>
<td>( \zeta_1 )</td>
<td>-0.0012</td>
<td>0.0701</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \eta_1 )</td>
<td>-0.0271</td>
<td>-0.1887</td>
<td>( \delta_1 )</td>
<td>0.0182</td>
<td>0.0529</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \nu_1 )</td>
<td>-0.0314</td>
<td>-0.026</td>
<td>( \delta_1 )</td>
<td>-0.0204</td>
<td>0.0147</td>
<td></td>
<td></td>
</tr>
<tr>
<td>( \phi_1 )</td>
<td>-0.0068</td>
<td>0.0962</td>
<td>( \phi_1 )</td>
<td>-0.0042</td>
<td>0.005</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 5 Pay-off table of PIS.

<table>
<thead>
<tr>
<th>( Y_1 )</th>
<th>( Y_2 )</th>
</tr>
</thead>
<tbody>
<tr>
<td>Max ( Y_1 )</td>
<td>1.02</td>
</tr>
<tr>
<td>Max ( Y_2 )</td>
<td>0.91</td>
</tr>
</tbody>
</table>
average thrust obtained by 2D FEM and the analytical model are 1017N and 1064N, respectively. So, the 4.62% error has been only occurred. Thrust ripples of the optimized motor, obtained from 2D FEM and the analytical model are respectively, 9.8% and 9.6% with 2.04% error. As a result, a good agreement between the analytical model and FEM analysis results is achieved, so it can be declared the effectiveness as well as the accuracy of the analytical model is verified too.

Now, considering the volume density of the iron weight as 7.86×10^4 N/m^3, the weight of the segmented secondary of the motor, which is lied along the rail are reduced from 92.45 N/m (9.245 kg) to 69.8 N/m (6.98 kg). This means 24.5% improvement (decrease) in

---

**Table 6 Design variables and objective values for the optimized motor.**

<table>
<thead>
<tr>
<th>Design Variables</th>
<th>Initial Motor</th>
<th>Optimized Motor</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Parameter [unit]</strong></td>
<td><strong>Initial Motor</strong></td>
<td><strong>Optimized Motor</strong></td>
</tr>
<tr>
<td>Field tooth width [mm]</td>
<td>15</td>
<td>38</td>
</tr>
<tr>
<td>Armature tooth width [mm]</td>
<td>15</td>
<td>16</td>
</tr>
<tr>
<td>Slot width [mm]</td>
<td>20</td>
<td>16</td>
</tr>
<tr>
<td>Slot height [mm]</td>
<td>70</td>
<td>40</td>
</tr>
<tr>
<td>Slot opening width [mm]</td>
<td>14</td>
<td>9</td>
</tr>
<tr>
<td>Slot opening height [mm]</td>
<td>7</td>
<td>2.5</td>
</tr>
<tr>
<td>Segment height [mm]</td>
<td>18</td>
<td>13</td>
</tr>
<tr>
<td>Segment back-factor</td>
<td>0.62</td>
<td>0.5</td>
</tr>
<tr>
<td>Segment factor</td>
<td>0.8</td>
<td>0.9</td>
</tr>
<tr>
<td>Tooth tip angle [deg]</td>
<td>35</td>
<td>23</td>
</tr>
<tr>
<td>SCLR</td>
<td>0.8</td>
<td>0.75</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Objective Values</th>
<th>Analytical</th>
<th>2D FEM</th>
<th>Analytical</th>
<th>2D FEM</th>
</tr>
</thead>
<tbody>
<tr>
<td>Average thrust [N]</td>
<td>599</td>
<td>578.5</td>
<td>1064</td>
<td>1017</td>
</tr>
<tr>
<td>Thrust-ripples [%]</td>
<td>78</td>
<td>77.6</td>
<td>9.6</td>
<td>9.8</td>
</tr>
</tbody>
</table>

Fig. 7 Initial motor modeled by 2D FEM a) motor structure and b) thrust force waveform.

Fig. 8 Optimized motor modeled by 2D FEM a) motor structure and b) thrust force waveform.
rail material consumption is achieved. However, considering the volume density of the copper weight as $9.8\times10^3$ N/m$^3$, the weight of the primary of the motor is increased from 332.5N (33.25 kg) to 357.8N (35.78 kg), i.e. 7.6%. Since the primary is a short type and considering the very good improvement in motor thrust and thrust ripples and material consumption in the segmented secondary, this increase is negligible. Considering the motor secondary only along its primary, the initial motor weight is 37.27 kg while the optimized motor weight is changed to 38.78 kg, i.e. 4% only increase in motor weight where the average ripple is increased 76% and the thrust ripple decreased sharply in comparing to initial motor.

6 Conclusion

A multi-level two-objective design optimization is performed on a SSLFSM to achieve high developed thrust and reduced thrust ripples. The motor dimensions are considered as design variables. According to the effect of the design variables on the optimization objectives, sensitivity analysis method is used to divide the design variables into two levels: as the mild-sensitive level and strong-sensitive level. Then, the two levels of design variables are optimized based on a mathematical model. The mathematical model is obtained by solving Maxwell and Poisson equations in certain regions of the motor and then, an analytical model is proposed for calculation of thrust and thrust ripples. Two different optimization methods are used; DOE and RSM in mild-sensitive level and GA in strong-sensitive level. As a result, dimensions of the optimized motor are carried out. 75.8% development in thrust and 87.4% development in thrust ripples are achieved. Also, the weight of the segmented secondary of the motor, which is lied along the rail are reduced from 92.45 N/m to 69.8 N/m with 24.5% improvement in rail material consumption. However, the weight of the primary of the motor is increased from 332.5N to 357.8N, i.e. 7.6%, but the primary is a short type and also considering the very good improvement in motor thrust and thrust ripples, this increase is negligible. A finite element analysis is employed to verify the mathematical modeling and optimization results. The accuracy of the mathematical model in thrust and thrust ripples calculations is verified with errors less than 5%. Finally, the effectiveness of the analytical modeling and the proposed optimization method is verified.

Acknowledgements

This work was supported by ALBORZ Province Power Distribution Company.

References


M. S. Hosseini received the B.Sc. and M.Sc. degrees in Electrical Engineering both from the School of Electrical and Computer Engineering, University of Tehran, Tehran, Iran, in 2005 and 2009 respectively. She is currently working toward the Ph.D. degree in Electrical Engineering in the School of Electrical Engineering, University of Shahid Beheshti, Tehran, Iran. Her research interests include design, modeling, and optimization of electrical machines and electromagnetic studies. She is also reviewer of IEEE Transactions on Magnetics, IEEE Transactions on Energy Conversion, IEEE Transactions on Vehicular Technology, IET Electric Power Applications and so on.

H. Javadi received his B.Sc. in Electrical Engineering from Sharif University of Technology (Iran) in 1980, his Master from UMIST (UK) in 1986, and his Ph.D. degree in 1994 from the INPT (France), then, he joined to Electrical Engineering department of Power and Water University of Technology (which is currently joined to Shahid Beheshti University (SBU)). He is currently an Associate Professor and is expert in the area of insulation and high voltage technology, electrical machines and as well as power system studies.

S. Vaez-Zadeh has been with the University of Tehran, since 1997. His research interests include renewable energy, electrical drives, contactless power transfer, and energy policy. He has co-authored more than 150 technical papers and holds one US patent. He is the author of Control of Permanent Magnet Synchronous Motors, OXFORD University Press, 2018. Professor Vaez-Zadeh is an Editor of IEEE Transactions on Energy Conversion and a Subject Editor of IET Renewable Power Generation. He has been active in IEEE sponsored conferences as the General Chair and Keynote Speaker, and received several awards for his technical contributions.