General Formulation to Investigate Scattering from Multilayer Lossy Inhomogeneous Metamaterial Planar Structures

M. Kiani*, A. Abdolali*

Abstract: This paper presents a general formulation to investigate the scattering from Multilayer Lossy Inhomogeneous Metamaterial Planar Structure (MLIMPS) with arbitrary number of layers and polarizations. First, the dominating differential equations of transverse components of electromagnetic fields in each layer are derived. Considering the general solution form of the differential equations and the boundary conditions of the problem; a set of linear equations is obtained. By solving these equations, the electromagnetic fields in all the layers and reflection and transmission coefficients are calculated. This method is used in analyzing two inhomogeneous planar layers consisting of conventional material and metamaterial in an interesting example. The presented results are useful for constructing general duality between conventional materials and metamaterials.

Keywords: Duality, General formulation, Multilayer Lossy Inhomogeneous Metamaterial Planar Structure (MLIMPS).

1 Introduction

Homogeneous planar structures are the most applicable structures in filters, shielding and etc. Inhomogeneous materials in these structures can optimize shielding effectiveness and provide less scattering and larger bandwidth [1-10]. There are different methods to analysis an inhomogeneous planar multilayer structure such as Taylor’s series expansion [11], Fourier’s series expansion [12], Equivalent source method [13], Method of moment [14], Inverse scattering method [15] and so on. All these methods are numerical and suggested for a slab composed of inhomogeneous medium. In this paper, a general formulation for investigating scattering from arbitrary Multilayer Lossy Inhomogeneous Planar Structure (MLIMPS) with an arbitrary number of layers and polarizations is presented. Dominating differential equations of MLIMPS are written in form of a linear second order differential equation with non-constant coefficient. General solution form of these equations is presented. Using boundary conditions of the problem, a set of linear equations is obtained which leads to calculating the electromagnetic fields in all layers and reflection and transmission coefficients. This formulation is applicable in stating general fundamentals and properties of multilayer inhomogeneous metamaterial structures and can be used for providing conditions for intended phenomena for the mentioned structures such as zero reflection, zero transmission, cloaking and etc. Finally, the presented method is applied in an interesting example that is useful for establishing a duality between conventional materials and metamaterials in general case [16].

2 Differential Equation of MLIMPS

Fig. 1 indicates a typical MLIMPS with N layers. The left half space is free space and the right one is an arbitrary homogeneous medium. Constitutive parameters of kth layer are as follow:

\[ \mu_k(z) = \mu_R(z) - j\mu_L(z) \]

\[ \varepsilon_k(z) = \varepsilon_R(z) - j\varepsilon_L(z) \]

for \( k = 1, 2, 3, ..., N \).

By combination of Maxwell’s Curl equations, we can write vector wave equations for kth layer as follows [17].

\[ \nabla \times (\mu_k(z)^{-1} \nabla \times E_k(z)) = \omega^2 \varepsilon_k(z) E_k(z) \]

\[ \nabla \times (\varepsilon_k(z)^{-1} \nabla \times H_k(z)) = \omega^2 \mu_k(z) H_k(z) \]

Because of the infinite dimension of the structure in y direction, we have:

\[ \frac{\partial}{\partial y} = 0 \]
Under phase-matching condition in the boundary of each two consecutive medium, we have

\[ Q(x, z) = P(z) e^{-jK_x x} \text{ or } \frac{\partial}{\partial x} = -jK_x \] (6)

where:

\[ K_x = K_0 \sin \theta, \]

\[ K_0 = \omega \varepsilon_0 \mu_0 \] (7)

The above equations, indicates the general form of electromagnetic field in each medium.

Using Eq. (3, 4) and noticing Eq. (5, 6), differential equation of transverse components of electromagnetic fields in kth medium can be obtained as follows:

For TEz polarization

\[ \frac{d^2 E_{ky}(z)}{dz^2} + \left( \omega^2 \mu_k(z)\varepsilon_k(z) - K_x^2 \right) E_{ky}(z) = 0 \] (9)

where:

\[ \mu_k(z) = \frac{d\mu_k(z)}{dz} \]

For TMz polarization

\[ \frac{d^2 H_{ky}(z)}{dz^2} - \left( \omega^2 \varepsilon_k(z)\mu_k(z) - K_x^2 \right) H_{ky}(z) = 0 \] (11)

where:

\[ \varepsilon_k(z) = \frac{d\varepsilon_k(z)}{dz} \]

Eqs. (9) and (11) are second order linear differential equations with non-constant coefficients. From principles of ordinary differential equations, second order linear differential equation of transverse fields in kth inhomogeneous medium has a general solution as follows:

\[ E_{ky}(z) = C_{k,1} f_k(z) + C_{k,2} g_k(z) \]

\[ H_{ky}(z) = C_{k,1} f_k(z) + C_{k,2} g_k(z) \] (14)

In Eq. (14), \( f_k \) and \( g_k \) show the variation of transverse electromagnetic fields in kth layer and \( C_{k,1} \) and \( C_{k,2} \) are unknown constant coefficients which can be calculated by using the boundary conditions named matching of tangential components of fields at the interface between kth and k+1th layer.

2.2 Boundary Conditions

Considering the general form of transverse fields in Eq. (14), the boundary conditions for tangential component of electric and magnetic fields in each consecutive layer are written as follows:

At the interface \( z = 0 \)

\[ F_{t1}(z) - F_{t1}(z)|_{z=0} = (C_{1,1} f_1(z) + C_{1,2} g_1(z))|_{z=0} = 0 \] (15)

\[ F_{t1}(z) - F_{t1}(z)|_{z=0} = 0 \]

\[ \frac{j}{\omega p_1(z)} (C_{1,1} f_1(z) + C_{1,2} g_1(z))|_{z=0} = 0 \]

\[ \frac{j}{\omega p_1(z)} (C_{1,1} f_1(z) + C_{1,2} g_1(z))|_{z=0} = 0 \]
At the interface \( z = d_k \) (\( k=1,2,\ldots,N-1 \))

\[
(C_{k,1}f_k(z) + C_{k,2}g_k(z))|_{z = d_k} = C_{k+1,1}f_{k+1}(z) + C_{k+1,2}g_{k+1}(z)
\]

and

\[
(C_{k,1}f_k(z) + C_{k,2}g_k(z))|_{z = d_k} = C_{k+1,1}f_{k+1}(z) + C_{k+1,2}g_{k+1}(z)
\]

Furthermore, at the interface \( z = d_N \)

\[
(C_{N,1}f_N(z) + C_{N,2}g_N(z))|_{z = d_N} = F^T_{1}(z)|_{z = d_N}
\]

and

\[
(C_{N,1}f_N(z) + C_{N,2}g_N(z))|_{z = d_N} = F^T_{1}(z)|_{z = d_N}
\]

In Eqs (15)-(20), we have the following definitions:

**Reflection and transmission coefficients are defined as follows:**

\[
\Gamma = \frac{F^t_1(0)}{F^t_1(0)}
\]

and

\[
T = \frac{F^T_{1}(d_N)}{F^T_{1}(0)}
\]

By substituting the Eqs. (30, 31) in Eqs. (15-29), 2N+2 linear equations with 2N+2 unknown variables are derived which can be solved easily to obtain reflection and transmission coefficients and electromagnetic fields in all regions. These unknown variables can be calculated as follows:

\[
\begin{cases}
C_{k,1}, C_{k,2} \\
1 \leq k \leq N
\end{cases}
\]

Therefore, a set of linear equations can be written as a suitable matrix of equations in the form of Eq. (33). This matrix of equations is applicable and so useful in analyzing the arbitrary type of MLIMPS and this formulation can be introduced as a full-wave method for the analysis of MILPS. As a potential application, this method can lead researchers to interesting phenomena such as zero reflection and zero transmission and can construct a novel duality between conventional materials and metamaterials in general case.
3 Example and Discussion

In this section, scattering of electromagnetic waves from two planar inhomogeneous structures composed of conventional material and metamaterial, situated between two free half spaces, are studied through an interesting example. Electromagnetic parameters of inhomogeneous layers are as shown in Table 1, where \( * \) denotes the complex-conjugate operator. It is assumed that \( \varepsilon_{r_0} = 4 \), \( K \) is an arbitrary complex constant and thickness of each inhomogeneous layer is \( d = 20 \) cm. A TE\(^z\) polarized plane wave with the electric field strength of \( E_i = 1.0 \) V/m and the excitation frequency of 1.0 GHz is illuminated to the assumed structures obliquely. According to relations purposed in [11-15], we can write general form of electric fields through Bessel functions \( J_v(\cdot) \) as follows:

\[
\begin{align*}
E_{y1}(x,z) &= e^{-jK_0 x} \\
[ A_1 ]^{J} \left( \frac{2K_0 \sqrt{\varepsilon r_0}}{K} \exp\left(\frac{-K}{2}z\right) \right) \\
+& B_1^{J} \left( \frac{2K_0 \sqrt{\varepsilon r_0}}{K} \exp\left(\frac{K}{2}z\right) \right)
\end{align*}
\]

(34)

According to differential equations concepts and Eq. (9) and the rules mentioned in [13] it can be easily concluded that considering electromagnetic parameters in these inhomogeneous slabs being negative conjugates of one another, general forms of variations in \( z \) direction in these structures will be conjugates of each other. Figs. 2-3 represent amplitudes and imaginary parts of transverse electric fields in the two structures at angles of incidence \( \theta = 60^\circ \) for \( K_1 = \) and \( K_5 = \). The purposed curves show great agreement with the results in [11-14].

Amplitudes and imaginary parts of reflection coefficients of the two structures at the angle of incidence between 0 and 90 for \( K_1 = \) and \( K_5 = \) are shown in Figs. 4-5. The results show great agreement with the curves purposed in [11-14].

As it can be seen from the purposed relations for the fields, they are not complex-conjugates of each other due to presence of exponential term \( e^{-jk_0 x} \) whereas reflection coefficients are conjugates of one another. It can be interpreted according to the concepts purposed in [16]. If we model inhomogeneous slabs of the two structures by two port networks and transfer matrixes as it is shown in Fig. (6), components of transfer matrices of these networks are conjugates of one another. Thus

\[
E_{y2}(x,z) = e^{-jK_0 x} \\
[ A_2 ]^{J} \left( \frac{2K_0 \sqrt{\varepsilon r_0}}{K} \exp\left(\frac{-K^*}{2}z\right) \right) \\
+ B_2^{J} \left( \frac{2K_0 \sqrt{\varepsilon r_0}}{K} \exp\left(\frac{K^*}{2}z\right) \right)
\]

(35)

Table 1 Constitutive parameters of the structures under consideration.

<table>
<thead>
<tr>
<th>Structure 1</th>
<th>Structure 2</th>
</tr>
</thead>
<tbody>
<tr>
<td>Relative permeability</td>
<td>( \mu_r = 1 )</td>
</tr>
<tr>
<td>Relative permittivity</td>
<td>( \mu_r = -1 )</td>
</tr>
</tbody>
</table>

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Fig. 2 Amplitudes of the transverse components of the electric fields related to structure 1 and structure 2.

Fig. 3 Imaginary parts of the transverse components of the electric fields for structure 1 and structure 2.

The impedances of source and load indicate the media of two half spaces. Since inhomogeneous slabs are situated between two half spaces composed of lossless media and according to the second theorem purposed in [16], considering definition of input impedance and reflection coefficients for two port networks, reflection coefficients of the two structures are conjugates of one another. Thus

\[
T_1 = \begin{bmatrix} A_1 & B_1 \\ C_1 & D_1 \end{bmatrix}
\]

(36)

\[
T_2 = \begin{bmatrix} A_2 & B_2 \\ C_2 & D_2 \end{bmatrix} = \begin{bmatrix} A_1^* & B_1^* \\ C_1^* & D_1^* \end{bmatrix}
\]

(37)

Fig. 4 Amplitude of the reflection coefficients for structure 1 and structure 2.

Fig. 5 Imaginary part of the reflection coefficients for structure 1 and structure 2.

Fig. 6 A two-port network model for inhomogeneous planar layer a) Structure 1 b) Structure 2.
\[ Z_{\text{in}}^1 = \frac{A_1 Z_{L1} + B_1}{C_1 Z_{L1} + D_1} \quad (38) \]
\[ Z_{\text{in}}^2 = \frac{A_2 Z_{L2} + B_2}{C_2 Z_{L2} + D_2} \quad (39) \]
\[ \Gamma_1 = \frac{Z_{\text{in}}^1 - Z_L}{Z_{\text{in}}^1 + Z_L} \quad (40) \]
\[ \Gamma_2 = \frac{Z_{\text{in}}^2 - Z_L}{Z_{\text{in}}^2 + Z_L} \quad (41) \]

Presented results can be seen as an initiative by researchers for establishing a duality between conventional and metamaterial media in much more general case than it is proved in [16].

5 Conclusion

General formulation to investigate the scattering from MLIMPS was presented. Using Maxwell’s equations and noticing physical and geometrical properties of the structure and plane wave which illuminate the assumed structure, dominating differential equations of MLIMPS were derived. From principles of ordinary differential equations, the general form of solution of wave equation for transverse components of electric and magnetic field is presented and with this general solution and using the boundary conditions of the problem, a set of linear equations was obtained. Solving these equations, reflection and transmission coefficient and electromagnetic fields in all regions were calculated. Finally the proposed method was used in an interesting example for two conventional material structure and metamaterial structure which can be useful for establishing general duality between conventional materials and metamaterials.

References


Mohammad Kiani was born in Amol, Iran on July 26, 1987. He received B.Sc. Degree from Ferdowsi University of Mashhad in 2010 and M.Sc. Degree from Iran University of Science and Technology in 2012, respectively, both in electrical engineering. His Scientific fields of interest are electromagnetic wave propagation and scattering from inhomogeneous media, complex media, electromagnetic compatibility and optimization problems.

Ali Abdolali was born in Tehran, Iran, on May 3, 1974. He received B.Sc. degree from the University of Tehran, and M.S. degree from the University of Tarbiat Modares, Tehran, and the Ph.D. degree from the Iran University of Science and Technology (IUST), Tehran Ann Arbor, all in electrical engineering, in 1998, 2000, and 2010, respectively. In 2010, he joined the Department of Electrical Engineering, Iran University of Science and Technology, Tehran, Iran, where he is an assistant Professor of electromagnetic engineering. His research interests include electromagnetic wave scattering, Radar Cross Section (RCS) & RCSR, Radar Absorbing Materials (RAM), cloaking, Metamaterials, Wave Propagation in composite media (anisotropic, inhomogeneous, dispersive media), Frequency Selective Surfaces (FSS), Bio electromagnetism (BEM). He has authored or coauthored over 50 papers in international journals & conferences.